Audio ICs

## Fluorescent display tube level meter driver, 16-point $\times 2$ channel, VU scale, bar display <br> BA6800AF / BA6805A

The BA6800AF and BA6805A are two-channel, 16-point fluorescent display tube drivers for VU-scale bar-level meters.
They use a dynamic-drive system and are provided with both $A C$ and $D C$ inputs. The $A C$ input mode has a peak hold circuit.
The ICs feature a power-on mute, and the output block can directly drive fluorescent display tubes, so few external components are required.
The grid output duty cycle is $1 / 8$ for the BA6800AF and $1 / 4$ tor the BA6805A. Apart from power dissipation, all other characteristics are the same.

- Applications

Level meters for all types of AV equipment
-Features

1) Uses dynamic-drive system to display two 16-point channels. Packages are 28-pin DIP (BA6805A) and 28-pin SOP (BA6800AF)
2) $A C$ and $D C$ inputs provided. Switching function allows two-mode display.
3) Upper 12 points have peak hold function in AC mode (two seconds).
4) Power-on mute function.
5) Dynamic-drive system reduces the power consumption of the fluorescent display tube power supply.
6) Square root compression amplifier built in
-Block diagram

|  |  |
| :---: | :---: |

- Absolute maximum ratings $\left(\mathrm{Ta}=25^{\circ} \mathrm{C}\right)$

| Parameter |  | Symbol | Limits | Unit |
| :---: | :---: | :---: | :---: | :---: |
| Supply voltage |  | $V_{C C}$ | 7.0 | V |
| Power dissipation | BA6805A | Pd | 700*1 | mW |
|  | BA6B00AF |  | 550*2 | mW |
| Operating temperature |  | Topr | $-20 \sim 70$ | ${ }^{\circ} \mathrm{C}$ |
| Storage temperature |  | Tstg | $-55 \sim 125$ | ${ }^{\circ} \mathrm{C}$ |
| Output voltage |  | $\mathrm{TcCl}^{\text {, }}+\mathrm{V}_{\text {EE }}$ | 36 | $\checkmark$ |

* 1 Reduced by 7 mW for each increase in Ta of $1^{\circ} \mathrm{C}$ over $25^{\circ} \mathrm{C}$.
$* 2$ Reduced by 5.5 mW for each increase in Ta of $1^{\circ} \mathrm{C}$ over $25^{\circ} \mathrm{C}$.
Electrical characteristics (unless otherwise specified $\mathrm{Ta}=25^{\circ} \mathrm{C}$ )

| Parameter |  | Symbol | Min. | Typ. | Max. | Unit | Conditions | Measurement Circuit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Supply vollage |  | $\mathrm{V}_{\mathrm{cc}}$ | 4.5 | 5.0 | 5.8 | $\checkmark$ |  | Fig. 1 |
| Quiescent current |  | la | - | 17 | 24 | mA | - | Fig. 1 |
| $A C$ input resistance |  | Rinac | 175 | 250 | 325 | Q | 1, Эpin | Fig. 1 |
| DC input resistance |  | Rivoc | 7 | 10 | 13 | $k \Omega$ | - | Fig. 1 |
| Oscillator frequency |  | foso | 1.7 | 2.0 | 2.3 | kHz | $\mathrm{C}=0.015 \mu \mathrm{~F}, \mathrm{R}=30 \mathrm{k} \Omega$ | Fig. 1 |
| Peak hold time |  | Thok | - | 2 | - | s | $\mathrm{fosc}=2 \mathrm{kHz}$ | Fig. 1 |
| Output duty cycle | BA6B00AF | Duty | - | 1/8 | - | - | $\mathrm{fosc}=2 \mathrm{kHz}$ | Fig. 1 |
|  | BA6805A |  | - | 1/4 | - | - | $\mathrm{fosc}=2 \mathrm{kHz}$ | Fig. 1 |
| Grid low-level outpuit voltage |  | $V_{\text {Gl }}$ | - | 0.4 | 0.8 | $V$ | $\mathrm{t}_{\mathrm{a}}=5 \mathrm{~mA}$ | Fig. 1 |
| Grid output leak current |  | $I_{G}$ leak | - | - | 10 | $\mu \mathrm{A}$ | $\mathrm{Vcc}_{\mathrm{cc}}=5 \mathrm{~V}$ | Fig. 1 |
| 690 |  |  |  | Mamin |  |  |  |  |

-Electrical characteristics (unless otherwise specified $\mathrm{Ta}=25^{\circ} \mathrm{C}$ )

| Parameter | Symbol | Min. | Typ. | Max. | Unit | Conditions | Measurement Circuit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Segment high-level output valtage | $\mathrm{VOH}_{\mathrm{OH}}$ | 3.7 | 4.0 | - | $\checkmark$ | $10=2 \mathrm{~mA}$ | Fig. 1 |
| Segment output leak current | loberk | - | - | 10 | $\mu \mathrm{A}$ | $-\mathrm{V}_{E E}=-31 \mathrm{~V}$ | Fig. 1 |
| Input switching threshold | $V_{\text {TH }}$ | 2.2 | 2.5 | 2.8 | V | AC: $\operatorname{pin} 7$ "H", DC: $\operatorname{pin} 7$ "L" | Fig. 1 |
| $A C$ sensitivity | Vinac | 250 | 400 | 630 | mV | $\begin{aligned} & \mathrm{RIN}=1.5 \mathrm{k} \Omega \\ & \mathrm{AC} \text { comparator } 10 \text { on tevel } \end{aligned}$ | Fig. 1 |
| Maximum grid output current | 1 cm | 5 | - | - | mA | V 人 $=0.8 \mathrm{~V}$ | Fig. 1 |
| Maximum segment output current | lom | 2 | - | - | mA | $\mathrm{V}_{\mathrm{OH}}=3.7 \mathrm{~V}$ | Fig. 1 |
| AC comparator level 16 | $V_{\text {ciasc }}$ | 8.5 | 10 | 12 | dB | Pin 11 output | Fig. 1 |
| AC comparator level 15 | $\mathrm{V}_{\text {C154C }}$ | 6.0 | 7 | 8.5 | dB | Pin 12 output | Fig. 1 |
| AC comparator level 14 | $\mathrm{V}_{\text {Giac }}$ | 4.0 | 5 | 6.0 | dB | Pin 13 output | Fig. 1 |
| AC comparator level 13 | $V_{\text {cisac }}$ | 2.5 | 3 | 4.0 | dB | Pin 14 output | Fig. 1 |
| AC comparator level 12 | $V_{\text {Cl2 }}$ | 1.5 | 2 | 2.5 | dB | Pin 15 output | Fig. 1 |
| AC comparator level 11 | $\mathrm{V}_{\text {cinac }}$ | 0.5 | 1 | 1.5 | dB | Pin 16 output | Fig. 1 |
| AC comparator level 10 | $V_{\text {cinac }}$ | - | 0 | - | dB | Pin 17 output | Fig. 1 |
| AC comparator level 9 | $V_{\text {cbas }}$ | -1.5 | -1 | -0.5 | dB | Pin 18 output | Fig. 1 |
| AC comparator level 8 | $V_{\text {caac }}$ | -2.5 | -2 | -1.5 | dB | Pin 19 output | Fig. 1 |
| AC comparator level 7 | $V_{\text {ciac }}$ | -4.0 | -3 | -2.5 | dB | Pin 20 output | Fig. 1 |
| AC comparator level 6 | $V_{\text {ctac }}$ | -6.0 | -5 | -4.0 | dB | Pin 21 output | Fig. 1 |
| AC comparator level 5 | Vcsac | -8.5 | -7 | -6.0 | dB | Pin 22 output | Fig. 1 |
| AC comparator level 4 | $V_{\text {canc }}$ | -15 | -10 | -8.5 | dB | Pin 23 output | Fig. 1 |
| AC comparator level 3 | V ${ }^{\text {csac }}$ | -25 | -20 | -15 | dB | Pin 24 output | Fig. 1 |
| AC comparator level 2 | $V_{\text {czac }}$ | -35 | -30 | -25 | dB | Pin 25 output | Fig. 1 |
| AC comparator level 1 | $V_{\text {ciac }}$ | -55 | -40 | -35 | dB | Pin 26 output | Fig. 1 |
| DC comparator level 16 | $V_{C 1000}$ | 2.76 | 3.10 | 3.44 | V | Pin 11 output | Fig. 1 |
| DC comparator level 15 | $\mathrm{V}_{\text {Cisde }}$ | 2.35 | 2.64 | 2.93 | $V$ | Pin 12 output | Fig. 1 |
| DC comparator level 14 | $\mathrm{V}_{\text {ciade }}$ | 2.07 | 2.33 | 2.59 | v | Pin 13 output | Fig. 1 |
| DC comparator level 13 | $\mathrm{V}_{\text {cisec }}$ | 1.86 | 2.10 | 2.34 | V | PIn 14 output | Fig. 1 |
| DC comparator level 12 | $\mathrm{V}_{\text {cizoc }}$ | 1.79 | 2.03 | 2.27 | v | Pin 15 output | Fig. 1 |
| DC comparator level 11 | $V_{\text {C110c }}$ | 1.62 | 1.88 | 2.14 | V | Pin 16 output | Fig. 1 |
| DC comparator level 10 | $\mathrm{V}_{\text {ciooc }}$ | 1.51 | 1.80 | 2.09 | V | Pin 17 output | Fig. 1 |
| DC comparator level 9 | Vcapc | 1.40 | 1.71 | 2.02 | V | Pin 18 output | Fig. 1 |
| DC comparator level B | $V_{\text {cad }}$ | 1.33 | 1.66 | 1.99 | $v$ | Pin 19 output | Fig. 1 |
| DC comparator level 7 | V V coc | 1.23 | 1.58 | 1.93 | V | Pin 20 output | Fig. 1 |
| DC comparator level 6 | $V_{\text {ceoc }}$ | 1.07 | 1.41 | 1.75 | V | PIn 21 output | Fig. 1 |
| DC comparator level 5 | $V_{\text {c } 50}$ | 0.93 | 1.26 | 1.59 | v | Pin 22 output | Fig. 1 |
| DC comparator level 4 | Vatac | 0.77 | 1.07 | 1.37 | V | Pin 23 output | Fig. 1 |
| DC comparator level 3 | $V_{\text {c30 }}$ | 0.42 | 0.60 | 0.78 | V | Pin 24 output | Fig. 1 |
| DC comparator level 2 | $\mathrm{V}_{\text {c2de }}$ | 0.21 | 0.33 | 0.45 | V | Pin 25 output | Fig. 1 |
| DC comparator level 1 | $V_{\text {CID }}$ | 0.11 | 0.18 | 0.25 | V | Pin 26 output | Fig. 1 |

(2) DC comparator level $\mathrm{VC}(\mathrm{n}) \mathrm{DC}$ (Max.) $>\mathrm{VC}(n+1) \mathrm{DC}$ (Min.), but when the nth comparator is off, the ( $n+1$ )th comparator is never on.

- Measurement circuit


Fig. 1
Application example


Fig. 2
-Circuit operation
(1) Input block

The $A C$ input pins are pins are 1 and 3 , and the $D C$ in put pins are pins 5 and 6 (for both the BA6800AF and BA6805A). Pin 7 is used to switch between the AC and DC inputs. When the input to pin 7 is " H ", $A C$ input is selected (pins 1 and 3). Therefore, by using pin 7 to switch between the AC and DC modes, the IC can do two jobs. For example, pins 1 and 3 can be used for audio signal input, and pins 5 and 6 can be used as the input for the signal meter output from a tuner (DC). The $A C$ input impedance of pins 1 and 3 is a low $250 \Omega$ (typ.), so connect potentiometers ( $\mathrm{VR}_{1}$ and $\mathrm{VR}_{2}$ ) in series with the inputs to adjust the sensitivity and ch1 and ch2 balance.
(2) Peak hold circuit

The BA6800AF and BA6805A have peak hold circuits that temporarily holds peak signal levels in $A C$ input mode.
The peak hold function can be used with the upper 12 points (5 to 16). The peak hold time depends on the oscillator frequency. It is 2 sec. (typ.) for an oscillator frequency of 2 kHz .
DC mode does not have a peak hold function
(3) Grid output

The pin 9 and 10 grid outputs are open-collector NPN transistors. The logic is active low (the fluorescent tube lights when the output is "L"), so connect two PNP transistors $Q_{1}$ and $Q_{z}$ as shown in the application example circuit to drive the fluorescent tubes (see Fig. 3).


Fig. 3
(4) Segment output block

Pins 11 to 26 are the segment outputs. The output circuits are open-collector PNP transistors. When grid 1 is "L"., the ch 1 level is output (pin 1 or 5 input level), and when grid 2 is " $L$ ", the ch2 level is output (pin 3 or 6 input level). Refer to Fig. 4.
(5) Grid and segment output timing chart. The grid and segment output timing for an oscillator frequency of 2 kHz is shown in Fig. 5.
(6) Attack and release times

The response characteristic for $A C$ input signals is set by resistor $R_{1}$ and capacitor $C_{3}$ for ch1 and resistor $R_{2}$ and capacitor $C_{4}$ for ch2 (pins 2 and 4). When $R_{1}=$ $47 \mathrm{k} \Omega$ and $\mathrm{C}_{3}=22 \mu \mathrm{~F}$, the attack time is about 4 ms , and the release time is about 1 sec . (same for ch2).

Attack time : Time for the voltage on pins 2 and 4 to rise from 1 V to 2.5 V when the input goes from no input to 2.6 Vrms , then back to no input.

Release time : Time for the voltage on pins 2 and 4 to fall from 2.5 V to 1 V when the input goes from 2.6 Vrms to no input.
(7) Oscillator frequency

The resistor $\mathrm{R}_{26}$ and capacitor $\mathrm{C}_{5}$ connected to pin 28 determine the oscillator frequency
The oscillator frequency (fosc) and grid output period (T) are related as follows :
$T(\mathrm{~ms})=16 /$ fosc $(\mathrm{kHz})$


## -Timing chart



Fig. 5 (b)

- External components (refer to "Circuit operation")
$C_{1}$ and $C_{2}$ : input coupling capacitors.
$\mathrm{VR}_{1}$ and $V R_{2}$ : $A C$ sensitivity adjustment and balance adjustment ( $3 \mathrm{k} \Omega$ recommended).
$C_{3}, R_{1}, C_{4}$ and $R_{2}$ : set the response characteristics with respect to the $A C$ input signal. In the example given, the attack time is about 4 mS and the release time is about 1 sec.
$\mathrm{R}_{3}$ : Pullup resistor for the input switching terminal (pin 7).
$R_{6}$ and $R_{7}$ : resistor for the grid leak current prath (licesk). Set so that laleak $\times R_{6}\left(R_{7}\right)<0.6 \mathrm{~V}$.
$R_{4}$ and $R_{5}$ : base bias resistors for $Q_{1}$ and $Q_{2}$.
Conditions for base bias current (le) flow are $\mathrm{V}_{\mathrm{cc}}=5 \mathrm{~V}$ and $V_{F}=0.6 \mathrm{~V}$ :
$\frac{\mathrm{R}_{4}}{\mathrm{R}_{6}}<\frac{5-0.6}{0.6}=7.3$
the base current is given by the following formula.
$\mathrm{IG}(\mathrm{mA}) \fallingdotseq \frac{5-0.6}{\mathrm{R}_{4}(\mathrm{k} \Omega)}-\frac{0.6}{\mathrm{R}_{6}(\mathrm{k} \Omega)}$
$\mathrm{l}_{\mathrm{B}}>\frac{\text { Fluorescent tube grid current/hfe }}{\text { hes }}$
Set resistors $A_{4}$ and $R_{6}$ ( $R_{5}$ and $R_{7}$ ) so that
$Q_{1}$ and $Q_{2}$ : grid output inverting transistors. Use transistors for which $V_{c E o}>V_{c c}+V_{E E}$.
R8 to $R_{25}$ : Resistors that reverse bias the segments and grid when the fluorescent tube is not lit.
The application example given is for general cases Select the resistors to suit the characteristics of the fluorescent tube used.
$\mathrm{C}_{5}$ and $\mathrm{R}_{26}$ : set the oscillator frequency.
Capacitor $\mathrm{C}_{5}$ should be a component with good temperature characteristics.
- Operation notes

1) Adjust the potentiometers $\mathrm{VR}_{1}$ and $\mathrm{VR}_{2}$ (connected to pins 1 and 3) to adjust the 0 dB input level and the dispersion of ch1 and ch2.
2) The temperature characteristic for the lighting limit for the 16th LED is shown in Fig. 6.
3) The external resistor R26 connected to the oscillaor (pin 28) should be in the range of $20 \mathrm{k} \Omega$ and $100 \mathrm{k} \Omega$.
If it is outside this range, oscillation may stop due to the influence of temperature (see Fig. 7)

Electrical characteristics curves


Fig. 6 16th point lighting limit supply voltage vs. ambient temperature output current


OUTPUT CURRENT: $\mathrm{I}_{\mathrm{O}}$ (mA)

Fig. 9 Grid low-level output vs.


Fig. 7 Value of external components for oscillator ( $\mathrm{VCO}=5.0 \mathrm{~V}$ )


Fig. 8 Segment high-leve output vs. output current

External dimensions (Unit: mm)


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