LMV101,LMV102,LMV105,LMV110

LMV101/102/105/110 Fixed-Gain Amplifiers



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LMV101/102/105/110 **Fixed-Gain Amplifiers General Description**

The LMV101/102/105/110 fixed-gain amplifier family integrates a rail-to-rail op amp, two internal gain-setting resistors and a V⁺/2 bias circuit into one ultra tiny package, SC70-5 or SOT23-5. Fixed inverting gains of -1, -2, -5, and -10 are available.

The core op amp in this series is an LMV321, which provides rail-to-rail output swing, excellent speed-power ratio, 1MHz bandwidth, and 1V/µs of slew rate with low supply current.

The LMV101/102/105/110 family reduces external component count. It is the most cost effective solution for applications where low voltage operation, low power consumption, space savings, and reliable performance are needed. It enables the design of small portable electronic devices, and allows the designer to place the device closer to the signal source to reduce noise pickup and increase signal integrity.

Features

(For 5V Supply, Typical Unless Otherwise Noted)

- Fixed inverting gain available -1,-2,-5,-10 DC gain accuracy @2.7V supply - LMV101/102/105 2% (typ) 6% (typ) — LMV110 SC70-5 & SOT23-5 Space saving packages Industrial temperature range -40°C to +85°C
- Low supply current Rail-to-Rail output swing
- Guaranteed 2.7V and 5V performance

Applications

- General purpose portable devices
- Mobile communications
- Battery powered electronics
- Active filters
- Microphone preamplifiers



130µA

Absolute Maximum Ratings (Note 1)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

ESD Tolerance (Note 2)					
Machine Model	200V				
Human Body Model	1500V				
Supply Voltage (V + - V ⁻)	5.5V				
Output Short Circuit to V +	(Note 3)				
Output Short Circuit to V -	(Note 4)				
Mounting Temperature					
Infrared or Convection (20 sec)	235°C				

Storage Temperature Range-65°C to 150°CJunction Temperature (T_J, max)150°C(Note 5)150°C

Operating Ratings (Note 1)

Supply Voltage	2.7V to 5.0V
Temperature Range	$-40^{\circ}C \le T_{J} \le 85^{\circ}C$
Thermal resistance (θ_{JA})	
5-pin SC70-5	478°C/W
5-pin SOT23-5	265°C/W

2.7V Electrical Characteristics

Unless otherwise specified, all limits guaranteed for $T_J = 25^{\circ}C$, V⁺ = 2.7V, V⁻ = 0V, V_O = V ⁺/2 and R_L > 1M Ω . Boldface limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Typ (Note 6)	Max (Note 7)	Units
Vo	Output Swing	$R_L = 10k\Omega$ to 1.35V	V+-0.01	V+-0.1	V
					min
			0.08	0.18	V
					max
Is	Supply Current		80	170	μA
					max
	DC Gain Accuracy	LMV101, Gain = -1	2	5	%
		LMV102, Gain = -2	2	5	%
		LMV105, Gain = -5	2	6	%
		LMV110, Gain = -10	6	12	%
GBW	-3dB Bandwidth	LMV101, Gain = -1,	1.6		MHz
		$R_{L} = 2k\Omega, C_{L} = 100pF$			
		LMV102, Gain = -2,	1.8		MHz
		$R_{L} = 2k\Omega, C_{L} = 100pF$			
		LMV105, Gain = -5,	0.8		MHz
		$R_{L} = 2k\Omega, C_{L} = 100pF$			
		LMV110, Gain = -10,	0.2		MHz
		$R_{L} = 2k\Omega, C_{L} = 100pF$			

5V Electrical Characteristics

Unless otherwise specified, all limits guaranteed for $T_J = 25^{\circ}C$, $V^+ = 5V$, $V^- = 0V$, $V_O = V^+/2$ and $R_L > 1M\Omega$. Boldface limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Typ (Note 6)	Max (Note 7)	Units
Vo	Output Swing	$R_{L} = 2k\Omega$ to 2.5V	V+-0.04	V+-0.3	V
				V+-0.4	min
			0.14	0.3	V
				0.4	max
		$R_{L} = 10k\Omega$ to 2.5V	V+-0.01	V+-0.1	V
				V+-0.2	min
			0.1	0.18	V
				0.28	max
l _o	Output Current	Sourcing, $V_{O} = 0V$	60	5	mA
					min
		Sinking, $V_{O} = 5V$	160	10	mA
					min

5V Electrical Characteristics (Continued)

Unless otherwise specified, all limits guaranteed for $T_J = 25^{\circ}C$, $V^+ = 5V$, $V^- = 0V$, $V_O = V^+/2$ and $R_L > 1M\Omega$. Boldface limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Typ (Note 6)	Max (Note 7)	Units
ls	Supply Current		130	250	μA
				350	max
	DC Gain Accuracy	LMV101, Gain = -1	3.5	5	%
		LMV102, Gain = -2	3.5	5	%
		LMV105, Gain = -5	3.5	6	%
		LMV110, Gain = -10	9.0	12	%
SR	Slew Rate	(Note 8)	1		V/µs
GBW	-3dB Bandwidth	LMV101, Gain = -1, R $_{L}$ = 2k Ω , C $_{L}$ = 100pF	1.6		MHz
		LMV102, Gain = -2, R $_{L}$ = 2k Ω , C $_{L}$ = 100pF	1.8		MHz
		LMV105, Gain = -5 , R _L = 2k Ω , C _L = 100pF	0.8		MHz
		LMV110, Gain = -10 , R _L = $2k\Omega$, C _L = $100pF$	0.2		MHz

Note 1: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not guaranteed. For guaranteed specifications and the test conditions, see the Electrical Characteristics.

Note 2: Human body model, $1.5k\Omega$ in series with 100pF. Machine model, 0Ω in series with 100pF.

Note 3: Shorting circuit output to $\mathsf{V}^{\mathsf{+}}$ will adversely affect reliability.

Note 4: Shorting circuit output to V⁻ will adversely affect reliability.

Note 5: The maximum power dissipation is a function of $T_{J(max)}$, θ_{JA} , and T_A . The maximum allowable power dissipation at any ambient temperature is P $_D = (T_{J(max)} - T_A)/\theta_{JA}$. All numbers apply for packages soldered directly into a PC board.

Note 6: Typical Values represent the most likely parametric norm.

Note 7: All limits are guaranteed by testing or statistical analysis.

Note 8: Number specified is the slower of the positive and negative slew rates.

Typical Performance Characteristics (Unless otherwise specified, $V_s = +5V$, single supply, $T_A = 25^{\circ}C$.)

Supply Current vs. Supply Voltage



Sourcing Current vs. Output Voltage



25°C.) (Continued)



Typical Performance Characteristics (Unless otherwise specified, V_s = +5V, single supply, T_A =

10

5

180

90

90

270

360

DS101234-28

10M

Phase Phase

Margin (°)

DS101234-25

Typical Performance Characteristics (Unless otherwise specified, $V_s = +5V$, single supply, $T_A = 25^{\circ}C$.) (Continued)

LMV102 Close Loop Frequency Response



LMV105 Close Loop Frequency Response



LMV102 Close Loop Frequency Response



LMV105 Close Loop Frequency Response



LMV110 Close Loop Frequency Response



LMV110 Close Loop Frequency Response



LMV101/102/105/110

LMV101/102/105/110



Inverting Large Signal Pulse Response LMV105



Inverting Small Signal Pulse Response LMV101



Inverting Large Signal Pulse Response LMV102



Inverting Large Signal Pulse Response LMV110



Inverting Small Signal Pulse Response LMV102



DS101234-38

Typical Performance Characteristics (Unless otherwise specified, $V_s = +5V$, single supply, $T_A = 25^{\circ}C$.) (Continued)

Inverting Small Signal Pulse Response LMV105

Slew Rate vs. Supply Voltage



Application Information

The LMV101/102/105/110 integrates a rail-to-rail op amp, two internal gain-setting resistors and a V⁺/2 bias circuit into one ultra tiny package, SC70-5 or SOT23-5. With its small footprint and reduced component count for gain stage, it enables the design of smaller portable electronic products, such as cellular phones, pagers, PDAs, PCMCIA cards, etc. In addition, the integration solution minimizes printed circuit board stray capacitance, and reduces the complexity of circuit design.

The core op amp of this family is National's LMV321.

1.0 Supply Bypassing

The application circuits in this datasheet do not show the power supply connections and the associated bypass capacitors for simplification. When the circuits are built, it is always required to have bypass capacitors. Ceramic disc capacitors $(0.1\mu F)$ or solid tantalum $(1\mu F)$ with short leads, and located close to the IC are usually necessary to prevent interstage coupling through the power supply internal impedance. Inadequate bypassing will manifest itself by a low frequency oscillation or by high frequency instabilities. Sometimes, a $10\mu F$ (or larger) capacitor is used to absorb

Inverting Small Signal Pulse Response LMV110



low frequency variations and a smaller $0.1\mu F$ disc is paralleled across it to prevent any high frequency feedback through the power supply lines.

2.0 Input Voltage Range

The input voltage should be within the supply rails. The ESD protection circuitry at the input of the device includes a diode between the input pin and the negative supply pin. Driving the input more than 0.6V (at 25°C) beyond the negative supply will turn on the diode and cause signal distortions. For applications that require sensing voltages beyond the negative rail, use the LMV111 with external gain setting resistors.

Application Information (Continued)

3.0 Capacitive Load Tolerance

The LMV101/102/105/110 can directly drive 200pF capacitive load with Vs = 5V at -1 gain configuration without oscillation. Direct capacitive loading reduces the phase margin of amplifiers. The combination of the amplifier's output impedance and the capacitive load induces phase lag. This results in either an underdamped pulse or oscillation. To drive a heavier capacitive load, a resistive isolation can be used as shown in *Figure 1*.



FIGURE 1. Resistive Isolation of a Heavy Capacitive Load

The isolation resistor R_{iso} and the C_L form a pole to increase stability by adding more phase margin to the overall system. The desired performance depends on the value of R_{iso} . The bigger the R_{iso} resistor value, the more stable V_{OUT} will be. *Figure 2* is an output waveform of *Figure 1* using 100 Ω for R_{iso} and 1000pF for C_L .





4.0 Phase Inverting AC Amplifier

A single supply phase inverting AC amplifier can be easily built with the LMV101/102/105/110 series (*Figure 3*). The output voltage is biased at mid-supply, and AC input signal is amplified by (R_2/R_1). Capacitor C_{1N} acts as an input AC coupling capacitor to block DC potentials. A capacitor of 0.1µF or larger can be used. The output can swing

rail-to-rail. To avoid output distortion, the peak-to-peak amplitude of the input AC signal should be less than $V_{\rm CC}(R$ $_1/R_2).$



FIGURE 3. Phase Inverting AC Amplifier

It is recommended that a small-valued capacitor be used across the feedback resistor (R_2) to eliminate stability problems, prevent peaking of the response, and limit the bandwidth of the circuit. This can also help to reduce high frequency noise and some other interference. (See *Figure 4*)



5.0 Microphone preamplifier

Most microphones have a low output voltage level. This output signal needs to be amplified so that it can feed the next stage with optimal level. *Figure 5* shows a microphone preamplifier circuit with the LMV110. This microphone preamplifier can provide 20dB gain. It can be implemented in PCs, PDAs, and mobile phones.

Input capacitor $C_{\rm IN}$ serves two important functions. First, it blocks any DC voltage from the previous stage to prevent the output from shifting to some unwanted DC level. This

DS101234-12 could cause the output to saturate when audio signal is applied at the input. Second, the C_{IN} and the 10k input resistor form a low pass filter to block any low frequency noise. The cut-off frequency of this low pass filter is given by,

$$= \frac{1}{2\pi R_1 C_{\rm IN}}$$

where $R_1 = 10k\Omega$ in LMV110. Output capacitor C_{OUT} is used to block the DC output from the next stage. R _{bias} is selected according to the microphone requirement.

Application Information (Continued)



FIGURE 5. Microphone Preamplifier with 20dB Gain

To improve power supply ripple rejection of the above microphone preamplifier, another capacitor and a pot can be connected to pin 1 as shown in *Figure 6*. The impedance of the two capacitors at audio frequencies are low. The R_{POT} can be adjusted so that the supply ripples injected through both the inverting input and the non-inverting input cancel each other at the output. If we ignore the impedance of the capacitors, we can select the pot value based on the following equation:

$$\frac{Z_{OUT} \parallel R_{POT}}{Z_{OUT} \parallel R_{POT} + R_{bias}} \times V_{ripple} \times G = \frac{1}{2} \times V_{ripple} \times (G + 1)$$

 Z_{OUT} is the output impedance of the microphone, and G is the gain of the preamplifier in absolute value.



FIGURE 6. Improved Ripple Rejection

6.0 Adjustable-Gain Amplifier

The LMV101/102/105/110 not only provides fixed gain of -1, -2, -5, and -10, it can also be configured for different gains by adding only one external resistor.

You can decrease the gain by putting a resistor in series with pin 1 (*Figure 7*). You can increase the gain by connecting a resistor from pin 1 to pin 3 (*Figure 8*).







FIGURE 8. Increased Gain

If you are using the LMV110 as a microphone preamplifier for an electret microphone (*Figure 5*), and the output impedance of the microphone is $1k\Omega$, then the gain of the preamplifier is

$$G = \frac{100k}{10k || R + 1k}$$

If we choose a small value for R, then we could get a preamplifier with a gain close to 100 (40dB), which is 10 times the gain provided by LMV110.



Connection Diagrams IN-R 1 V_{OUT} 100k 100k DS101234-2 5 5-Pin SC70-5 (M7) DS101234-3 5-Pin SOT23-5 (M5) **Ordering Information** NSC Package Part number Marking DC Gain **R1 R2 Transport Media** Drawing LMV101M7 1k Units Tape and Reel A38 -1 100k 100k LMV101M7X 3k Units Tape and Reel LMV102M7 1k Units Tape and Reel A39 -2 100k 200k 3k Units Tape and Reel LMV102M7X SC70-5 MAA05A LMV105M7 1k Units Tape and Reel A40 -5 50k 250k LMV105M7X 3k Units Tape and Reel LMV110M7 1k Units Tape and Reel A41 -10 10k 100k LMV110M7X 3k Units Tape and Reel LMV101M5 1k Units Tape and Reel A33A 100k 100k -1 LMV101M5X 3k Units Tape and Reel LMV102M5 1k Units Tape and Reel A34A -2 100k 200k LMV102M5X 3k Units Tape and Reel SOT23-5 MA05B LMV105M5 1k Units Tape and Reel 250k A35A -5 50k LMV105M5X 3k Units Tape and Reel LMV110M5 1k Units Tape and Reel A36A -10 10k 100k LMV110M5X 3k Units Tape and Reel



LMV101/102/105/110



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