

SANYO Semiconductors DATA SHEET

An ON Semiconductor Company

Overview

The STK433-890N-E is 4 channels class-AB audio frequency power amplifier hybrid IC.

Application

• Audio Power amplifiers

Features

- Pin-to-pin compatible outputs ranging from 40W to 80W.
- Output load impedance: $R_L = 6\Omega$ recommended.
- Allows the use of predesigned applications for standby and mute circuit.

Series model

	STK433-040N-E	STK433-060N-E	STK433-130N-E	STK433-330N-E
Output1 (10%/1kHz)	$40W \times 2ch$	50W imes 2ch	150W imes 2ch	150W imes 3ch
Output2 (0.4%/20Hz to 20kHz)	$25W\times2ch$	35W imes 2ch	$100W \times 2ch$	$100W\times 3ch$
Max. rating V _{CC} (quiescent)	±38V	±46V	±71.5V	±71.5V
Max. rating V _{CC} (6 Ω)	±36V	±40V	±63V	±63V
Recommended operating V _{CC} (6 Ω)	±24V	±27V	±44V	±44V
Dimensions (excluding pin height)	47.0mm×25.	6mm×9.0mm	67.0mm×25.6mm×9.0mm	64.0mm×36.6mm×9.0mm

	STK433-840N-E	STK433-870N-E	STK433-890N-E
Output1 (10%/1kHz)	$40W \times 4ch$	$60W \times 4ch$	$80W \times 4ch$
Output2 (0.4%/20Hz to 20kHz)	25W imes 4ch	$40W \times 4ch$	50W imes 4ch
Max. rating V _{CC} (quiescent)	±38V	±50V	±54V
Max. rating V_{CC} (6 Ω)	±36V	±44V	±47V
Recommended operating V _{CC} (6 Ω)	±25V	±30V	±34V
Dimensions (excluding pin height)	64.0mm×31.	78.0mm×44.1mm×9.0mm	

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• Allowable load shorted time: 0.3 second

• Miniature package.

Specifications

Absolute Maximum Ratings at $Ta = 25^{\circ}C$, $Tc = 25^{\circ}C$ unless otherwise specified

Parameter	Symbol	Conditions	Ratings	Unit
Maximum power supply voltage	V _{CC} max (0)	Non signal	±54	V
	V _{CC} max (1)	Signal, $R_L \ge 6\Omega$	±47	V
	V _{CC} max (2)	Signal, $R_L = 4\Omega$	±40	V
Minimum operation supply voltage	V _{CC} min		±10	V
#13 Operating voltage *5	VST OFF max	#13pin voltage	-0.3 to +5.5	V
Thermal resistance	өј-с	Per power transistor	2.1	°C/W
Junction temperature	Tj max	Both the Tj max and Tc max	150	°C
Operating substrate temperature	Tc max	conditions must be met.	125	°C
Storage temperature	Tstg		-30 to +125	°C
Allowable time for load short-circuit *4	ts	$V_{CC} = \pm 34V$, $R_L = 6\Omega$, f = 50Hz P _O = 50W, 1ch drive	0.3	S

Operating Characteristics at $Tc = 25^{\circ}C$, $R_{L} = 6\Omega$ (Non-inductive Load), $Rg = 600\Omega$, VG = 30dB

			C	onditions '							
Parameter	Symbol	V _{CC} [V]	f [Hz]	PO [W]	THD [%]		min	typ	max	Unit	
Output power *1	P _O 1	±34	20 to 20k		0.6		47	50			
	P _O 2	±34	1k		10			80		W	
Total harmonic distortion *1	THD 1	±34	20 to 20k						0.6		
	THD 2	±34	1k	5.0		VG=30dB		0.02		%	
Frequency characteristics *1	fL, fH	±34		1.0		+0 -3dB		20 to 50k		Hz	
Input impedance	ri	±34	1k	1.0				55		kΩ	
Output noise voltage *3	V _{NO}	±40				Rg=2.2kΩ			1.0	mVrms	
Quiescent current	Icco	±40				No load	90	150	210	mA	
Quiescent current at stand-by	ICST	±40				VST=0V			1.0	mA	
Neutral voltage	V _N	±40					-70	0	+70	mV	
#13 Stand-by ON threshold *5	VST ON	±34				Stand-by		0	0.6	V	
#13 Stand-by OFF threshold *5	VST OFF	±34				Operation	2.5	3.0	5.5	V	

Note

*1. 1channel operation.

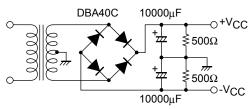
- *2. All tests are measured using a constant-voltage supply unless otherwise specified
- *3. The output noise voltage is peak value of an average-reading meter with a rms value scale (VTVM).

A regulated AC supply (50Hz) should be used to eliminate the effects of AC primary line flicker noise

- *4. Allowable time for load short-circuit and output noise voltage are measured using the specified transformer power supply.
- *5. The impression voltage of '#13 (Stand-By) pin' must not exceed the maximum rating. Power amplifier operate by impressing voltage +2.5 to +5.5V to '#13 (Stand-By) pin'.
- * Please connect $PreV_{CC}$ pin (#1 pin) with the stable minimum voltage.
- and connect so that current does not flow in by reverse bias.
- * In case of heat sink design, we request customer to design in the condition to have assumed market.
- * The case of this Hybrid-IC is using thermosetting silicon adhesive (TSE322SX).
- * Weight of HIC : (typ) 37.0g

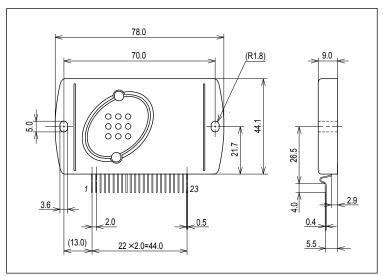
Outer carton dimensions (W×L×H) : $452mm \times 325mm \times 192mm$

Specified transformer power supply (Equivalent to MG-200)



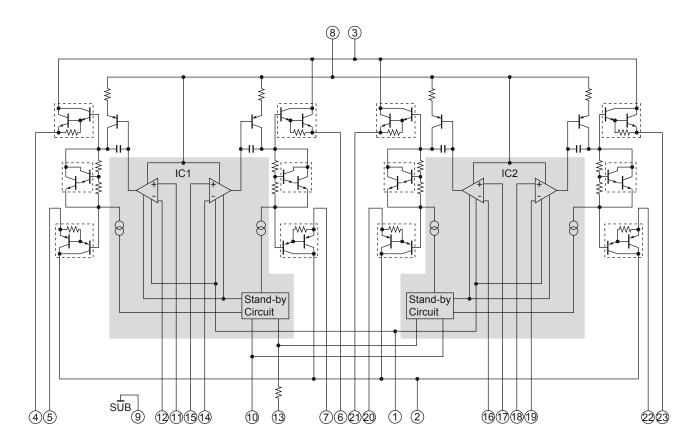
Package Dimensions

unit : mm (typ)

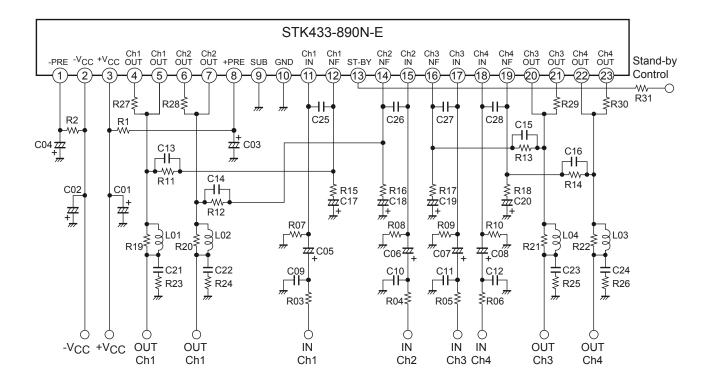


RoHS directive pass

Equivalent Circuit

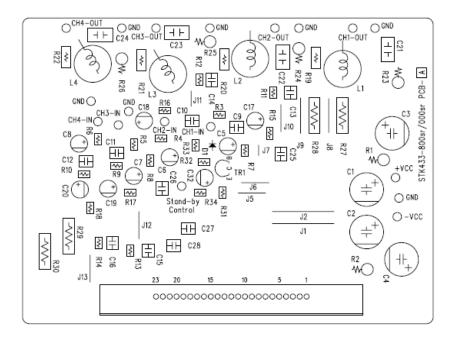


Application Circuit

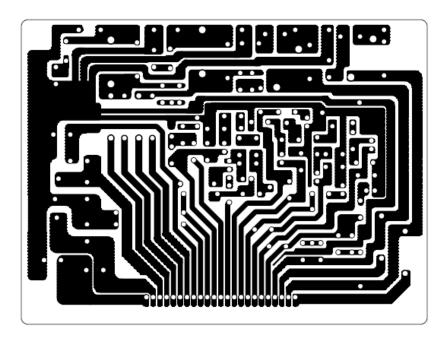


PCB Layout Example

Top view



Bottom view



STK433-800NSr PCB PARTS LIST

PCB Name : STK403-800Sr PCBA

	tion No.	RATING	Component								
LUCA	uon no.	RATING	STK433-840N-E/890N-E	STK433-870N-E							
Hybrid IC#1 Pin Posi	tion	-	1								
R01, R02		100Ω, 1W	0								
R03, R04, R05, R06		1kΩ, 1/6W	0								
R07, R08, R09, R10,	R11, R12, R13, R14	56KΩ, 1/6W	0								
R15, R16, R17, R18		1.8KΩ, 1/6W	0								
R19, R20, R21, R22		4.7Ω, 1/4W	0								
R23, R24, R25, R26		4.7Ω, 1W	0								
R27, R28, R29, R30		0.22Ω, 5W	0								
R32, R33, R34, R35		0.22Ω, 5W	-	0							
C01, C02, C03, C04		100μF, 100V ο									
C05, C06, C07, C08		2.2μF, 50V	o (*1)					2.2μF, 50V			
C09, C10, C11, C12		470pF, 50V	F, 50V o								
C13, C14, C15, C16		***pF, 50V	3pF								
C17, C18, C19, C20		10μF, 10V	• ((*1)							
C21, C22, C23, C24		0.1µF, 50V	0								
C25, C26, C27, C28		***pF, 50V	100p	F							
L01, L02, L03, L04		ЗμН	0								
	Tr1	VCE \geq 50V, IC \geq 10mA	0								
	D1	Di	0								
Stand-By	R31	1.8kΩ, 1/6W	0 0								
Control	R32	33kΩ, 1/6W									
Circuit	R33	1kΩ, 1/6W	0								
	R34	2kΩ, 1/6W	0								
	C32	33μF, 10V	0								
		-									
		-									
		-									
		-									

(*1) Capacitor mark "A" side is " – " (negative).

Recommended external components

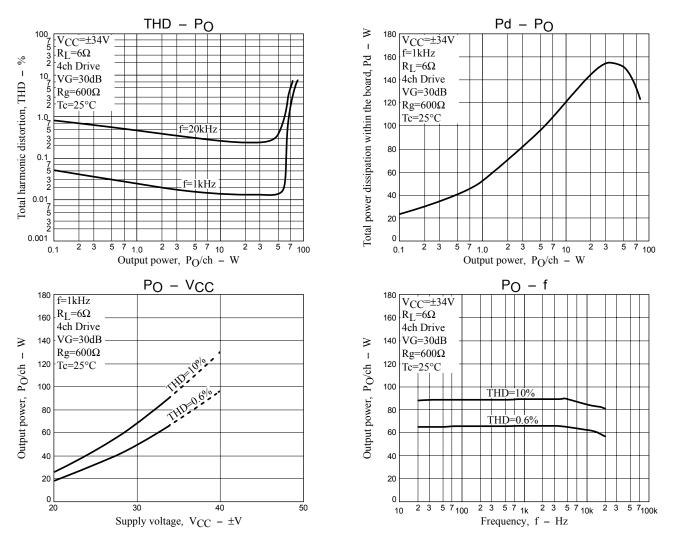
STK433-840N-E/890N-E

Parts	Recommended		Above	Below						
Location	value	Circuit purpose	Recommended value							
R01, R02	100Ω/1W	Resistance for Ripple filters. (Fuse resistance is recommended.	Recommended value Short-through current	Short-through current						
- , -		Ripple filter is constituted with C03, C04.)	may decrease at	may increase at high						
		· · · · · · · · · · · · · · · · · · ·								
R03,R04,R05,	1kΩ	Resistance for input filters.		frequency.						
R06			-	-						
R07,R08,R09,	56kΩ	Input impedance is determined.	Output neutral voltage	(VN) shift.						
R10		······································	(It is referred that R07=R11, R08=R12,							
			R09=R13, R10=R14)	,						
R11,R12,R13,	56kΩ	Voltage Gain (VG) is determined with R15, R16, R17, R18								
R14			-	-						
R15,R16,R17,	1.8kΩ	Voltage Gain (VG) is determined with R11, R12, R13, and R14.	It may oscillate.	With especially no						
R18	1.0122	(As for VG, it is desirable to set up by R15, R16, R17, and R18.)	(Vg < 30dB)	problem						
R19,R20,R21,	4.7Ω	Resistance for oscillation prevention.	(19 0002)	p.00.000						
R22	4.7 22		-	-						
R23,R24,R25,	4.7Ω/1W	Resistance for oscillation prevention.								
R26	4.732/100		-	-						
R27,R28,R29,	0.22Ω	Output emitter resi@stor (Metal-plate Resistor is recommended.)	Decrease of	It may cause thermal						
R30	±10%, 5W			runaway						
130	±10%, 3W		Maximum output	Tullaway						
R31	Noto *1	Select Destriction registered, for the improvement voltage of (#12	Power	t avaaad the maximum						
RUI	Note *4	Select Restriction resistance, for the impression voltage of '#13	(Stand-By) pin must no							
001 000	400	rating.								
C01, C02	100μF/100V	Capacitor for oscillation prevention.								
		Locate near the HIC as much as possible.	-	-						
		• Power supply impedance is lowered and stable operation of								
		the IC is carried out. (Electrolytic capacitor is recommended.)								
C03,C04	100μF/100V	Decoupling capacitor	The change in the Ripp	-						
		The Ripple ingredient mixed in an input side Is removed from a	an input side from a por	wer supply line						
005 000 007	0.0	power supply line. (Ripple filter is constituted with R01, R02.)								
C05,C06,C07, C08	2.2μF/50V	Input coupling capacitor. (For DC current prevention.)		-						
	470-5	lanut filter conceitor								
C09,C10,C11,	470pF	Input filter capacitor								
C12		• A high frequency noise is reduced with the filter constituted by		-						
012 014 015	EnE	R03, R04, R05, R06.	It may appliete							
C13,C14,C15,	5pF	Capacitor for oscillation prevention.	It may oscillate.							
C16	10E/10\/	Negative feedback conseilor		The voltage gain (1/2)						
C17,C18,C19,	10μF/10V	Negative feedback capacitor.	The voltage gain (VG)	The voltage gain (VG)						
C20		The cutoff frequency of a low cycle changes.	of low frequency is	of low frequency						
		$(fL = 1/(2\pi \cdot C17 \cdot R15))$	extended. However,	decreases.						
			the pop noise at the							
			time of a power							
			supply injection also							
			becomes large.							
C21,C22,C23,	0.1µF	Capacitor for oscillation prevention.	It may oscillate.							
C24										
C25,C26,C27,	100pF	Capacitor for oscillation prevention.	It may oscillate.							
C28										
L01,L02,L03,	3μΗ	Coil for oscillation prevention.	With especially	It may oscillate.						
L04			no problem							

[STK433-000N/-100N/-800)Nsr Pin Layout]																						
	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15								
(Size) 47.0mm×25.6mm×9.0mm						2c	h clas	sAB/	2.00r	nm													
STK433-040N 40W/JEITA	-	-	+	0	0	0	0	+			Ι	Ν	s	Ν	Ι								
STK433-060N 50W/JEITA	Р	V	V	U	U	U	U	Ρ	s	G	Ν	F	Т	F	Ν								
	R	C	C	Т	Т	Т	Т	R	U	N	/	/	A	/	/								
	E	С	С	C	C	, C	, C	E	В	D	С Н	С Н	N D	С Н	С Н								
(Size) 67.0mm×25.6mm×9.0mm				н	н	н	н				1	1		2	2								
STK433-130N 150W/JEITA				1	1	2	2						В										
				+	-	+	-						Υ										
	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20	21	22	23
(Size) 64.0mm×31.1mm×9.0mm										4c	h clas	sAB/	2.00r	nm									
STK433-840N 40W/JEITA	-	-	+	0	0	0	0	+			Ι	Ν	S	Ν	Ι	Ν	Ι	Ι	Ν	0	0	0	0
STK433-870N 60W/JEITA	Р	V	V	U	U	U	U	Р	S	G	Ν	F	т	F	Ν	F	Ν	Ν	F	U	U	U	U
	R	С	С	Т	Т	Т	Т	R	U	Ν	/	/	А	/	/	/	/	/	/	Т	Т	Т	Т
	E	С	С	/ C	/ C	/ C	/ C	Е	В	D	С Н	С Н	N D	С Н	С Н	С Н	С Н	С Н	С Н	/ C	/ C	/ C	/ C
(Size) 78.0mm×44.1mm×9.0mm				н	н	н	н				п 1	п 1		п 2	п 2	п 3	п 3	п 4	п 4	н	н	н	н
STK433-890N 80W/JEITA				1	1	2	2						В	-	-	Ũ	Ũ			3	3	4	4
				+	-	+	-						Υ							-	+	-	+
	1																						
	1									I		I			I								

[STK433-000N/-100N/-800Nsr Pin Layout]

Characteristic of Evaluation Board



A Thermal Design Tip For STK433-890N-E Amplifier

[Thermal Design Conditions]
The thermal resistance (θ c-a) of the heat-sink which manages the heat dissipation inside the Hybrid IC will be
determined as follow:
(Condition 1) The case temperature (Tc) of the Hybrid IC should not exceed 125°C
$Pd \times \theta c - a + Ta < 125^{\circ}C \cdots (1)$
Where Ta : the ambient temperature for the system
(Condition 2) The junction temperature of each power transistor should not exceed 150°C
$Pd \times \theta c - a + Pd/N \times \theta j - c + Ta < 150^{\circ}C^{\circ}$
Where N : the number of transistors (two for 1 channel, ten for channel)
θ j-c : the thermal resistance of each transistor (see specification)
Note that the power consumption of each power transistor is assumed to be equal to the total power dissipation (Pd)
divided by the number of transistors (N).
From the formula (1) and (2), we will obtain:
$\theta c-a < (125 - Ta)/Pd$ (1)
$\theta c-a < (150 - Ta)/Pd - \theta j-c/N$ (2)
The value which satisfies above formula (1)' and (2)' will be the thermal resistance for a desired heat-sink.
Note that all of the component except power transistors employed in the Hybrid IC comply with above conditions.
[Example of Thermal Design]
Generally, the power consumption of actual music signals are being estimated by the continuous signal of
$1/8 P_{O}$ max. (Note that the value of $1/8 P_{O}$ max may be varied from the country to country.)
(Sample of STK433-890N-E; 50W×4ch)
If V _{CC} is $\pm 34V$, and R _L is 6 Ω , then the total power dissipation (Pd) of inside Hybrid IC is as follow;
Pd = 99.0W (at 6.25W output power, 1/8 of P _O max)
There are eight (8) transistors in Audio Section of this Hybrid IC, and thermal resistance (θ j-c) of each transistor is
2.1°C/W. If the ambient temperature (Ta) is guaranteed for 50°C, then the thermal resistance (θ c-a) of a desired heat-
sink should be;
From (1)' $\theta c - a < (125 - 50)/99.0$
< 0.76
$From (2)^2$, $0 = 0 \neq (150, -50)/(00, 0) = 2, 1/8$

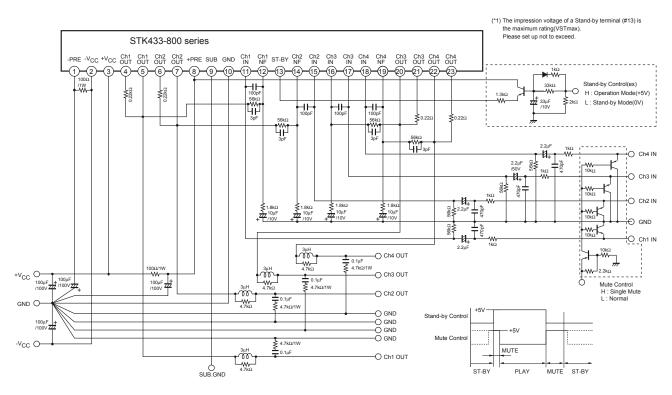
From (2)' $\theta c - a < (150 - 50)/99.0 - 2.1/8$ < 0.75

Therefore, in order to satisfy both (1)' and (2)', the thermal resistance of a desired Heat-sink will be 0.75°C/W.

[Note]

Above are reference only. The samples are operated with a constant power supply. Please verify the conditions when your system is actually implemented.

STK433-800 series Stand-by Control & Mute Control & Load-Short Protection Application

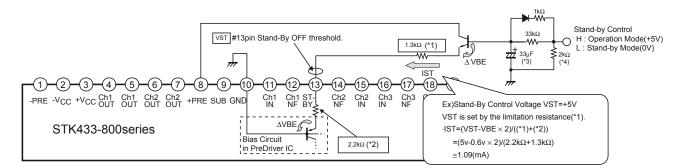


[STK433-800 series Stand-By Control Example]

[Feature]

- The pop noise generated when power supply ON/OFF by using recommendation Stand-By Control Application can be improved.
- Stand-By Control can be done by additionally adjusting the limitation resistance (*1) to the voltage such as Micro computer, the set design is easy.

(Reference circuit) STK433-800 series test circuit To Stand-By Control added +5V.



[Operation explanation]

1) #13pin Stand-By Control Voltage VST

(1) Operation Mode SW transistor of Stand-By Circuit is turned on when VST ≥ 2.5V or more is impressed, and the power amplifier works.
ex) VST = 3.0V VST = (*2) × IST+0.6V → 3.0V = 2.2kΩ × IST+0.6V Therefore, <u>IST≅1.09mA</u>
(2) Stand-By Mode VST ≤ 0.6V or less turns off the SW transistor of Stand-By Circuit by (typ 0V), and the amplifier stops.
ex) VST = 0.6V VST = (*2) × IST+0.6V → 0.6V = 2.2kΩ × IST+0.6V Therefore, <u>IST≅0mA</u>

- (*3) When the power supply is turned on by giving the time constant with the capacitor (*3) when the amplifier works, the pop noise is improved.
- (*4) When capacitor (*3) is discharged when the amplifier operation stops, the constant is decided.

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